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# PIN PHOTODETECTOR



The high electric field present in the depletion region causes photo-generated carriers to separate and be collected across the reverse -biased junction. This give rise to a current flow in an external circuit, known as photocurrent.

#### ENERGY-BAND DIAGRAM FOR A PIN PHOTODIODE



# PHOTOCURRENT

Optical power absorbed, in the depletion region can be written in terms of incident optical power:

$$P(x) = P_0(1 - e^{-\alpha_s(\lambda)x})$$

Absorption coefficient strongly depends on wavelength. The upper wavelength cutoff for any semiconductor can be determined by its energy gap as follows:

$$\lambda_c(\mu \mathrm{m}) = \frac{1.24}{E_g(\mathrm{eV})}$$

Taking entrance face reflectivity into consideration, the absorbed power in the width of depletion (megion, w, becomes:  $(1-R_f)P(w) = P_0(1-e^{-\alpha_f})(1-R_f)$ ).

Quantum Efficiency ( $\eta$ ) = number of e-h pairs generated / number of incident photons

Responsivity measures the input-output gain of a detector system. In the specific case of a photodetector, responsivity measures the electrical output per optical input.

$$\Re = \frac{I_p}{P_0} = \frac{\eta q}{h\nu} \quad \text{mA/mW}$$

### RESPONSIVITY



When  $\lambda << \lambda_c$  absorption is low When  $\lambda > \lambda_{c:}$  no absorption



### LIGHT ABSORPTION COEFFICIENT

The upper cutoff wavelength is determined by the bandgap energy Eg of the material.

At lower-wavelength end, the photo response diminishes due to low absorption (very large values of  $\alpha$ s).



# PHOTODETECTOR NOISE

In fiber optic communication systems, the photodiode is generally required to detect very weak optical signals.

Detection of weak optical signals requires that the photodetector and its amplification circuitry be optimized to maintain a given signal-to-noise ratio.

The power signal-to-noise ratio S/N (also designated by SNR) at the output of an optical receiver is defined by

$$SNR = \frac{S}{N} = \frac{\text{signal power from photocurrent}}{\text{photodetector noise power + amplifier noise power}}$$
SNR Can NOT be improved by amplification

# QUANTUM (SHOT NOISE)

Quantum noise arises due optical power fluctuation because light is made up of discrete number of photons

$$\left\langle i_{Q}^{2}\right\rangle = 2qI_{p}BM^{2}F(M)$$

F(M): APD Noise Figure F(M)  $\sim = Mx (0 \le x \le 1)$ I<sub>p</sub>: Mean Detected Current B = Bandwidth q: Charge of an electron

### DARK/LEAKAGE CURRENT NOISE

There will be some (dark and leakage ) current without any incident light. This current generates two types of noise

**Bulk Dark Current Noise** 

$$\langle i_{DB}^2 \rangle = 2qI_D BM^2 F(M)$$
  
I<sub>D</sub>: Dark Current

Surface Leakage Current Noise (not multiplied by M)

$$\left\langle i_{DS}^{2}\right\rangle = 2qI_{L}B$$

I<sub>L</sub>: Leakage Current

# THERMAL NOISE

The photodetector load resistor  $R_L$  contributes to thermal (Johnson) noise current

$$\left\langle i_T^2 \right\rangle = 4K_B T B / R_L$$

 $K_{\text{B}}$ : Boltzmann's constant = 1.38054 X 10(-23) J/K T is the absolute Temperature

 Quantum and Thermal are the significant noise mechanisms in all optical receivers

• RIN (Relative Intensity Noise) will also appear in analog links

# SIGNAL TO NOISE RATIO

Detected current = AC 
$$(i_p)$$
 + DC  
 $(I_p)$   
Signal Power =  $\langle i_p^2 \rangle M^2$   
 $\langle i_p^2 \rangle M^2$   
 $SNR = \frac{2q(I_p + I_D)M^2F(M)B + 2qI_LB + 4k_BTB/R_L}{2q(I_p + I_D)M^2F(M)B + 2qI_LB + 4k_BTB/R_L}$ 

Typically not all the noise terms will have equal weight. Often thermal and quantum noise are the most significant.

#### RESPONSE TIME IN *PIN* PHOTODIODE



Transit time, t<sub>d</sub> and carrier drift velocity v<sub>d</sub> are related by

 $t_d = w / v_d$  For a high speed Si PD,  $t_d = 0.1$  ns

#### RISE AND FALL TIMES



Photodiode has uneven rise and fall times depending on:

1. Absorption coefficient  $\alpha_s(\lambda)$  and

2. Junction Capacitance C<sub>i</sub>

$$C_{j} = \frac{\mathcal{E}_{o}\mathcal{E}_{r}A}{W}$$

# JUNCTION CAPACITANCE

 $C_{j} = \frac{\mathcal{E}_{o}\mathcal{E}_{r}A}{4},$ 

 $\epsilon_o = 8.8542 \text{ x } 10(-12) \text{ F/m}$ ; free space permittivity  $\epsilon_r =$  the semiconductor dielectric constant

A = the diffusion layer (photo sensitive)

area

Large area photov detector the aver letter junction capacitance hence small bandwidth (low speed)

 $\rightarrow$  A concern in free space optical receivers

#### VARIOUS PULSE RESPONSES



Pulse response is a complex function of absorption coefficient and junction capacitance

#### COMPARISONS OF *PIN* PHOTODIODES

Parameter	Symbol	Unit	Si	Ge	InGaAs
Wavelength range	λ	nm	400-1100	800–1650	1100-1700
Responsivity	${\mathscr R}$	A/W	0.4-0.6	0.4-0.5	0.75-0.95
Dark current	$I_D$	nA	1–10	50-500	0.5-2.0
Rise time	τ,	ns	0.5-1	0.1-0.5	0.05-0.5
Modulation (bandwidth)	$B_m$	GHz	0.3-0.7	0.5-3	1-2
Bias voltage	$V_B$	V	5	5–10	5

#### BASIC PIN PHOTODIODE STRUCT



**Rear Illuminated Photodiode** 

# PIN DIODE STRUCTURES



Diffused structures tend to have lower dark current than mesa etched structures although they are more difficult to integrate with electronic devices because an additional high temperature processing step is required.

